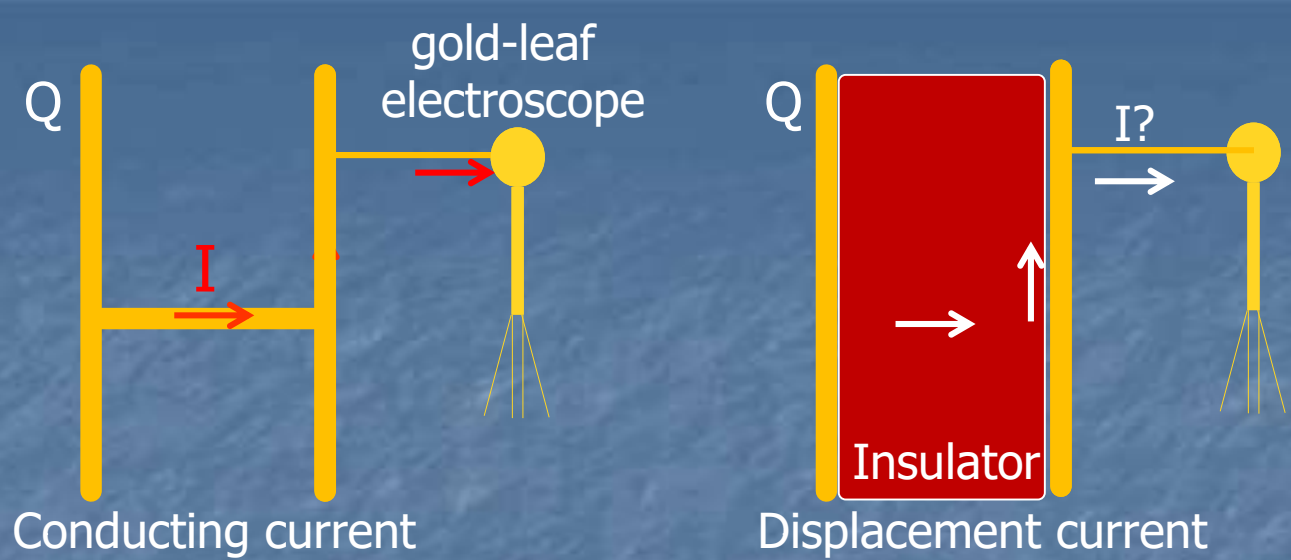


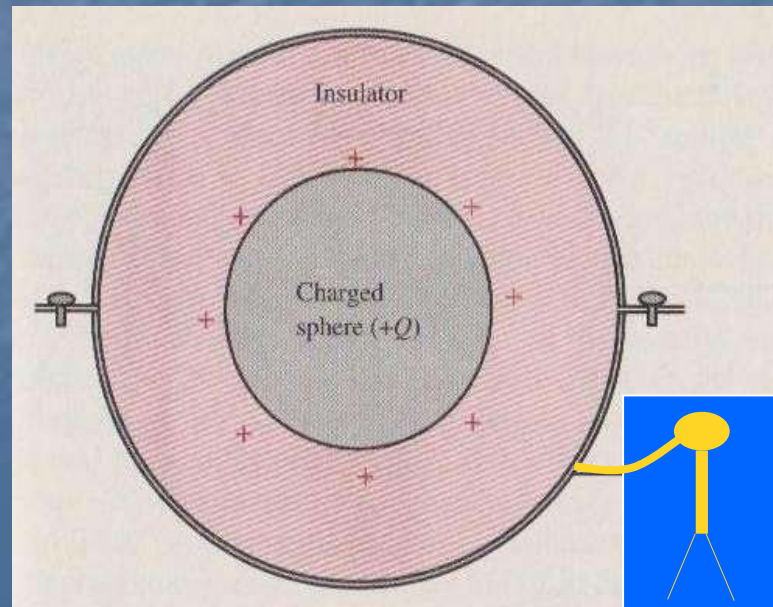
Examples

History of the displacement vector D

Michael Faraday
(1791 –1867)



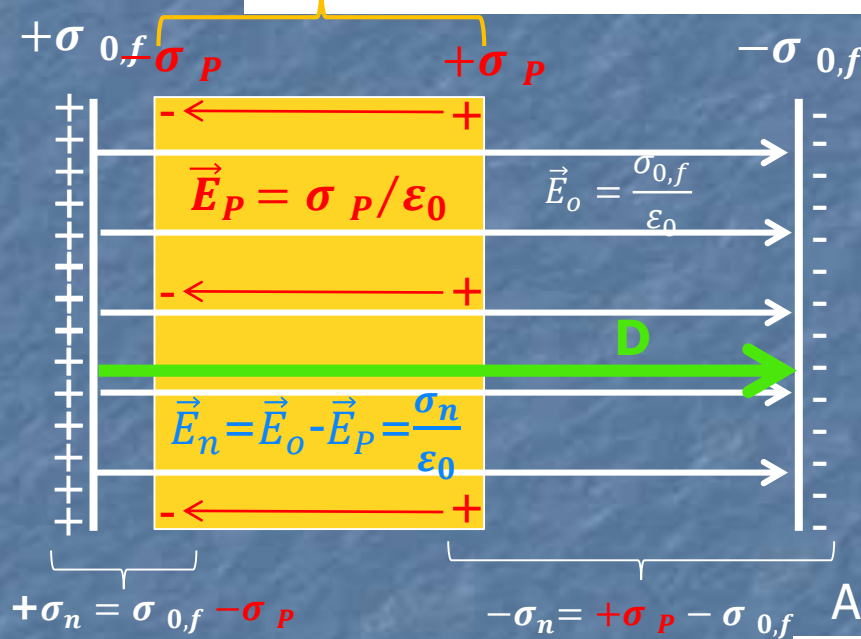
- Faraday tried many types of insulators but never prevented the displacement of charge
- The charge displaced was always Q , equal to that placed on the left plate.
- A sketch of the original system used by Faraday:



The relation between $\epsilon_0, \epsilon, \epsilon_r, \vec{D}, \vec{E}$ in Linear-Isotropic-Homogeneous medium

$$\oint_S \vec{E} \cdot d\vec{s} = \oint_S E_N ds = \frac{q_0}{\epsilon_0} \times \epsilon_0 \Rightarrow \oint_S \epsilon_0 E_N ds = q_0 \rightarrow$$

$$\rightarrow \epsilon_0 E_N \equiv D_N \quad (\epsilon_0 \vec{E} = \vec{D}) \Rightarrow \oint_S D_N ds = q_0 = \oint_S \vec{D} \cdot d\vec{s} \Rightarrow \text{div} \vec{D} = \rho_0$$



The polarization per unit volume:

$$P = \frac{p}{V} = \frac{Q_P \cdot d}{V} = \frac{\sigma_P \cdot A \cdot d}{V} = \frac{\sigma_P \cdot V}{V} = \sigma_P \quad [1]$$

The relation between \vec{E} and \vec{P} is:

$$\vec{P} = \epsilon_0 \chi_e \vec{E} \quad [2]$$

$$\vec{D} = \epsilon_0 \vec{E}_0 = \epsilon_1 \vec{E}_1 = \epsilon_i \vec{E}_i \quad ; \quad \left\{ \frac{q_o}{S} \right\} = \{ \sigma_o \} \quad [3]$$

And therefore the resultant field in the medium is:

$$E_N = E = E_0 - E_p = \left(E_0 - \frac{\sigma_p}{\epsilon_0} \right) / \epsilon_0 \Rightarrow$$

$$\Rightarrow \epsilon_0 E = \epsilon_0 E_0 - \sigma_p = \epsilon_0 E_0 - P = D - P = D - \epsilon_0 \chi_e E \Rightarrow$$

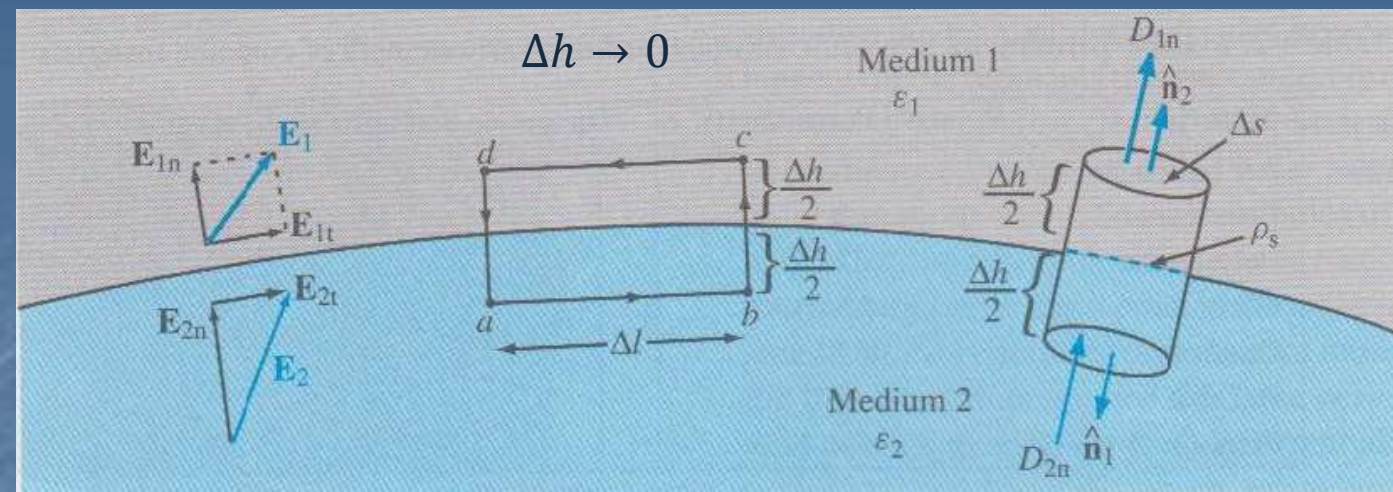
$$\Rightarrow D = \epsilon_0 E + \epsilon_0 \chi_e E = \epsilon_0 (1 + \chi_e) E \equiv \epsilon E \Rightarrow (1 + \chi_e) = \frac{\epsilon}{\epsilon_0} \equiv \epsilon_r > 1$$

unitless

A privet case of orthogonal fields:

Boundary condition for the electric field

Is this true in the case of time dependent fields?



$$1. \quad 0 = \oint_c \vec{E} \cdot d\vec{l} = \int_a^b E_{1T} dl + \int_c^d -E_{2T} dl \Rightarrow E_{1T} = E_{2T} \Rightarrow \frac{D_{1T}}{\epsilon_1} = \frac{D_{2T}}{\epsilon_2}$$

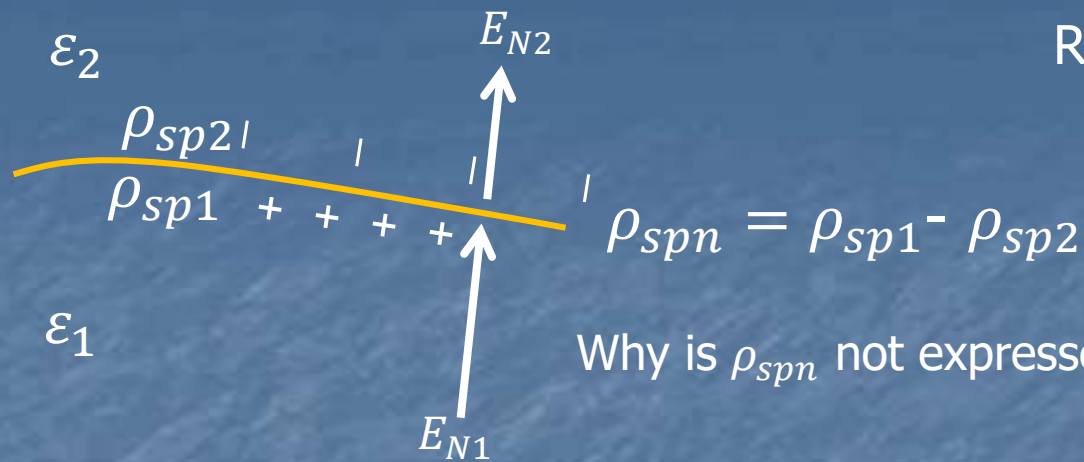
$$2. \quad Q_s = \oint_s \vec{D} \cdot d\vec{s} = \int_{top} D_{1N} ds + \int_{bottom} -D_{2N} ds = (D_{1N} - D_{2N}) \Delta s \Rightarrow \frac{Q_s}{\Delta s} = \rho_s = D_{1N} - D_{2N} \Rightarrow \epsilon_1 E_{1N} - \epsilon_2 E_{2N} = \rho_s$$

When $\rho_s = 0 \Rightarrow (3) D_{1N} = D_{2N}$ and (4) $\epsilon_1 E_{1N} = \epsilon_2 E_{2N}$

An important emphasis:

ρ_s is not ρ_{sp}

Regarding the charge on the interface plan:



Why is ρ_{spn} not expressed in Equation 4?

When $\rho_s = 0$, (4) $\boxed{\varepsilon_1 E_{N1} = \varepsilon_2 E_{N2}}$; introducing $\varepsilon = \varepsilon_0 \varepsilon_r$:

$\varepsilon_0 \varepsilon_{r1} E_{N1} = \varepsilon_0 \varepsilon_{r2} E_{N2}$; introducing $\varepsilon_r = 1 + \chi$:

$\varepsilon_0 (1 + \chi_1) E_{N1} = \varepsilon_0 (1 + \chi_2) E_{N2} \Rightarrow$

$\varepsilon_0 E_{N1} + \varepsilon_0 \chi_1 E_{N1} = \varepsilon_0 E_{N2} + \varepsilon_0 \chi_2 E_{N2}$; recalling that $P = \varepsilon_0 \chi E$:

$\varepsilon_0 E_{N1} + P_{N1} = \varepsilon_0 E_{N2} + P_{N2}$; recalling that $P_N = \rho_{sp}$:

$\Rightarrow \varepsilon_0 (E_{N1} - E_{N2}) = P_{N2} - P_{N1} = \rho_{sp2} - \rho_{sp1}$ and therefore :

$\boxed{E_{N1} - E_{N2} = \frac{\rho_{sp2} - \rho_{sp1}}{\varepsilon_0} = \frac{\rho_{spn}}{\varepsilon_0}} \quad (5) \quad \text{As in a capacitor}$

The advantage of Eq. 4 over 5 is due to the value of ε being known, while ρ_{spn} is unknown and to be calculated.

The particular case of E_N

The electric fields of point charge, infinite wire, charged sphere, infinite charged plane and more, where all calculated for vector field perpendicular to the medium surface. Assuming medium-vacuum interface, one get from Eq. 4:

$$(4) \quad \varepsilon_1 E_{1N} = \varepsilon_2 E_{2N} \Rightarrow \varepsilon_0 E_0 = \varepsilon E \Rightarrow E = \frac{\varepsilon_0}{\varepsilon} E_0 = \frac{E_0}{\varepsilon_r}; \quad \boxed{E = \frac{E_0}{\varepsilon_r}}$$

That is to say that the field in vacuum is ε_r times greater than that in medium. Employed this on the case of electric field of a point charge one gets:

$$E = \frac{E_0}{\varepsilon_r} = \frac{\frac{q}{4\pi\varepsilon_0 r^2}}{\varepsilon_r} = \frac{q}{4\pi\varepsilon_0 \varepsilon_r r^2} = \frac{q}{4\pi\varepsilon r^2}$$

and for homogeneously charged infinite plane:

$$E = \frac{E_0}{\varepsilon_r} = \frac{\frac{\rho_s}{\varepsilon_0}}{\varepsilon_r} = \frac{\rho_s}{\varepsilon_r \varepsilon_0} = \frac{\rho_s}{\varepsilon}$$

and so on... and therefore in all the expressions of electric field one should introduces ε instead of ε_0 .

Is crossing the interface between two media might change the direction of the electric field vector and how?

As for the electric potential:

$$\frac{V}{V_0} = \frac{-\int_a^b \vec{E} \cdot d\vec{l}}{-\int_a^b \vec{E}_0 \cdot d\vec{l}} = \frac{-\int_a^b \frac{\vec{E}_0}{\varepsilon_r} \cdot d\vec{l}}{-\int_a^b \vec{E}_0 \cdot d\vec{l}} = \frac{1}{\varepsilon_r} \Rightarrow V = \frac{V_0}{\varepsilon_r}$$

As for the electric flux:

$$\frac{\phi}{\phi_0} = \frac{\oint_s \vec{E} \cdot d\vec{s}}{\oint_s \vec{E}_0 \cdot d\vec{s}} = \frac{\oint_s \frac{\vec{E}_0}{\epsilon_r} \cdot d\vec{s}}{\oint_s \vec{E}_0 \cdot d\vec{s}} = \frac{1}{\epsilon_r} \Rightarrow \boxed{\phi = \frac{\phi_0}{\epsilon_r}}$$

As for the effective charge q^* :

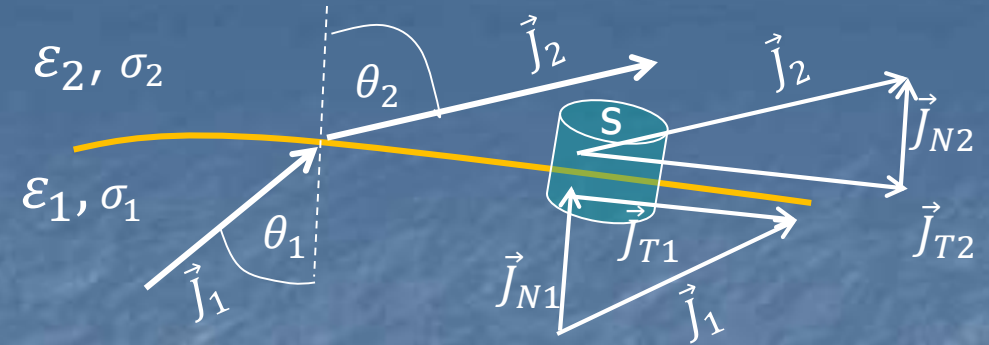
$$\frac{q^*}{\epsilon_0} = \phi = \frac{\phi_0}{\epsilon_r} = \frac{q/\epsilon_0}{\epsilon_r} = \frac{q}{\epsilon_0 \epsilon_r} = \frac{q}{\epsilon} \Rightarrow \boxed{q^* = \frac{\epsilon_0 q}{\epsilon} = \frac{q}{\epsilon_r}}$$

And the capacitance:

$$C = \frac{Q}{V} = \frac{Q}{V_0/\epsilon_r} = \boxed{\epsilon_r C_0 = C}$$

Continuity of the current density \vec{J}

What happens to \vec{J} in the interface conductor-insulator ($\sigma_2 \rightarrow 0$)?



From Kirchhoff Junction law we get:

$$0 = \oint_{\text{Cylinder}} \vec{J} \cdot d\vec{s} = \oint_{\text{Cy}} J_N ds = \int_{\text{top}} J_{N1} ds - \int_{\text{bottom}} J_{N2} ds \Rightarrow (J_{N1} - J_{N2}) \Rightarrow J_{N1} = J_{N2}$$

As for \vec{J}_T ? On that we learn from the continuity of \vec{E}_T

$$E_{T1} = E_{T2} \Rightarrow \frac{J_{T1}}{\sigma_1} = \frac{J_{T2}}{\sigma_2}$$

In insulator $J = 0 \Rightarrow J_{N1} = 0 = J_{N2}$. On the other hand $E_{N2} \neq 0$, yet contradicts the fact that $J_{N2} = 0$ since $J_{N2} = \sigma_2 E_{N2}$ but $\sigma = 0$. In any case, at any point within a conductor $E = D = 0$, i.e. $E_1 = D_1 = 0$, out of which $E_{T2} = 0$ and $D_{N2} = 0$

The relation between the direction of J_1 and J_2 at the interface:

$$\tan \theta_1 = \frac{J_{T1}}{J_{N1}}; \quad \tan \theta_2 = \frac{J_{T2}}{J_{N2}} = \frac{J_{T1} \frac{\sigma_2}{\sigma_1}}{J_{N1}} = \tan \theta_1 \frac{\sigma_2}{\sigma_1} \Rightarrow$$

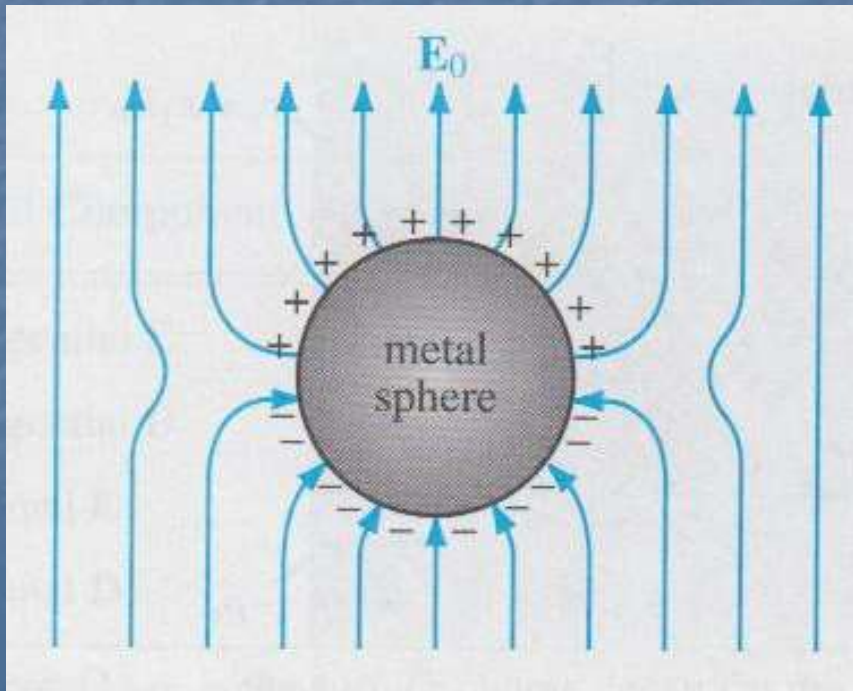
$$\tan \theta_2 = \tan \theta_1 \frac{\sigma_2}{\sigma_1}$$

Field Component	Any Two Media	Medium 1 Dielectric ϵ_1	Medium 2 Dielectric ϵ_2	Medium 1 Dielectric ϵ_1	Medium 2 Conductor
Tangential E	$E_{1t} = E_{2t}$	$E_{1t} = E_{2t}$		$E_{1t} = E_{2t} = 0$	
Tangential D	$D_{1t}/\epsilon_1 = D_{2t}/\epsilon_2$	$D_{1t}/\epsilon_1 = D_{2t}/\epsilon_2$		$D_{1t} = D_{2t} = 0$	
Normal E	$\hat{\mathbf{n}} \cdot (\epsilon_1 \mathbf{E}_1 - \epsilon_2 \mathbf{E}_2) = \rho_s$	$\epsilon_1 E_{1n} - \epsilon_2 E_{2n} = \rho_s$		$E_{1n} = \rho_s/\epsilon_1$	$E_{2n} = 0$
Normal D	$\hat{\mathbf{n}} \cdot (\mathbf{D}_1 - \mathbf{D}_2) = \rho_s$	$D_{1n} - D_{2n} = \rho_s$		$D_{1n} = \rho_s$	$D_{2n} = 0$

$$\tan \theta_1 = \frac{J_{T1}}{J_{N1}}; \quad \tan \theta_2 = \frac{J_{T2}}{J_{N2}} = \frac{J_{T1} \frac{\sigma_2}{\sigma_1}}{J_{N1}} = \tan \theta_1 \frac{\sigma_2}{\sigma_1} \Rightarrow$$

$$\tan \theta_2 = \tan \theta_1 \frac{\sigma_2}{\sigma_1}$$

Therefore, for $\sigma_1 \gg \sigma_2 \Rightarrow \theta_2 \rightarrow \frac{\pi}{2}$ and hence for every current arriving the conducting-insulator, it apparently flow along the interface. However, though current can not exist in insulator, electric field does.



In summary, the only component of the electric field exist is perpendicular to the interface conductor-insulator (see figure)

Does this result correlate with what we know about currents and equi-potential surfaces?

The continuity of J in the interface between two conductors:

From the continuity of D ($\Delta D_N = \rho_s$):

$$\epsilon_1 E_{1N} - \epsilon_2 E_{2N} = \rho_s ; \text{introducing } E = \frac{J}{\sigma}$$

$$\epsilon_1 \frac{J_{1N}}{\sigma_1} - \epsilon_2 \frac{J_{2N}}{\sigma_2} = \rho_s ; \text{since } J_{1N} = J_{2N} = J_N$$

$$\rho_s = J_N \left(\frac{\epsilon_1}{\sigma_1} - \frac{\epsilon_2}{\sigma_2} \right) = J_N \left(\frac{1}{\tau_1} - \frac{1}{\tau_2} \right)$$

In good conductors $\epsilon \rightarrow \epsilon_0$ and hence:

$$\rho_s = \epsilon_0 J_N \left(\frac{1}{\sigma_1} - \frac{1}{\sigma_2} \right)$$

Why a voltage difference is developed between the two sides of a resistor?

The duality between J and D (electrostatics):

Conductor	Insulator
$\oint \vec{E} \cdot d\vec{l} = 0$	$\oint \vec{E} \cdot d\vec{l} = 0$
$J = \sigma E = -\sigma \text{ grad} V$	$D = \epsilon E = -\epsilon \text{ grad} V$
$\text{div } J = 0$	$\text{Div } D = 0$
$\text{delta } J_N = 0$	$\text{delta } D_N = 0$
$J_{T1}/\sigma_1 = J_{T2}/\sigma_2$	$D_{1T}/\epsilon_1 = D_{T2}/\epsilon_2$
$R = l/\sigma A$	$C = \epsilon A/l$
$RC = \frac{\epsilon}{\sigma} = \tau$	

MKS :

$$\{\phi_B\} = \textit{Weber} (Wb)$$

$$\{\vec{B}\} = \frac{\phi_B}{m^2} = \frac{Wb}{m^2} \equiv 1T (\textit{Tesla})$$

$$1T = 10^4 \textit{Gauss} (cgs)$$

$$|\vec{B}_{Earth}| \approx 0.5G$$

$$\{Wb\} = \frac{Kg \cdot m^2}{sec^2 \cdot A} = \textit{Volt} \cdot \textit{sec} = T \cdot m^2 = \frac{J}{A} = 10^8 Mx$$

Biot–Savart law and the vector potential \vec{A}

$$\vec{B} = \frac{\mu_0 I}{4\pi} \oint_l \frac{d\vec{l} \times \hat{r}}{r^2} = \frac{\mu_0 I}{4\pi} \oint_l d\vec{l} \times \left[-\text{grad} \left(\frac{1}{r} \right) \right] \quad [1]$$

From vector analysis : $\text{curl}(\phi \vec{A}) = \phi \text{curl} \vec{A} - \vec{A} \times \text{grad} \phi \Rightarrow \vec{A} \times \text{grad} \phi = \phi \text{curl} \vec{A} - \text{curl}(\phi \vec{A})$ [2]

$$\vec{B} = -\frac{\mu_0 I}{4\pi} \oint_l \left[\frac{1}{r} \text{curl} d\vec{l} - \text{curl} \left(\frac{d\vec{l}}{r} \right) \right] = \oint_l \frac{\mu_0 I}{4\pi} \text{curl} \left(\frac{d\vec{l}}{r} \right) =$$

$$\text{curl} \oint_l \frac{\mu_0 I}{4\pi} \left(\frac{d\vec{l}}{r} \right) \equiv \text{curl} \vec{A} ; \quad \boxed{\vec{B} = \text{curl} \vec{A}} \quad [3]$$

$$\vec{A} \equiv \oint_l \frac{\mu_0 I}{4\pi} \left(\frac{d\vec{l}}{r} \right) = \oint_{l,s} \frac{\mu_0 J ds d\vec{l}}{4\pi r} = \oint_{l,s} \frac{\mu_0 J \hat{l} ds dl}{4\pi r} = \boxed{\iiint_V \frac{\mu_0 \vec{J} dv}{4\pi r}}$$

$$V = \frac{Q}{4\pi\epsilon r} = \boxed{\iiint_V \frac{\rho_v dv}{4\pi\epsilon r}}$$

consequently one get the similarity between V and \vec{A} :

$$\nabla^2 V = -\frac{\rho_v}{\epsilon} \Rightarrow \nabla^2 \vec{A} = -\mu \vec{J} \Rightarrow$$

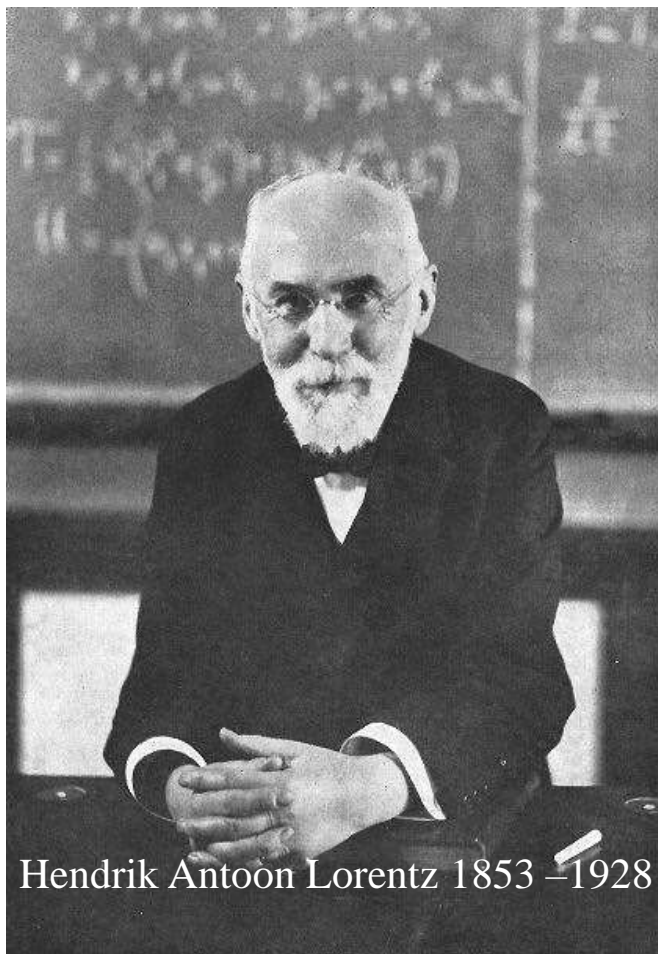
$$\nabla^2 A_x = -\mu J_x \ ; \ \nabla^2 A_y = -\mu J_y \ ; \ \nabla^2 A_z = -\mu J_z$$

The integral and differential expression of Ampere's law:

Stokes theorem

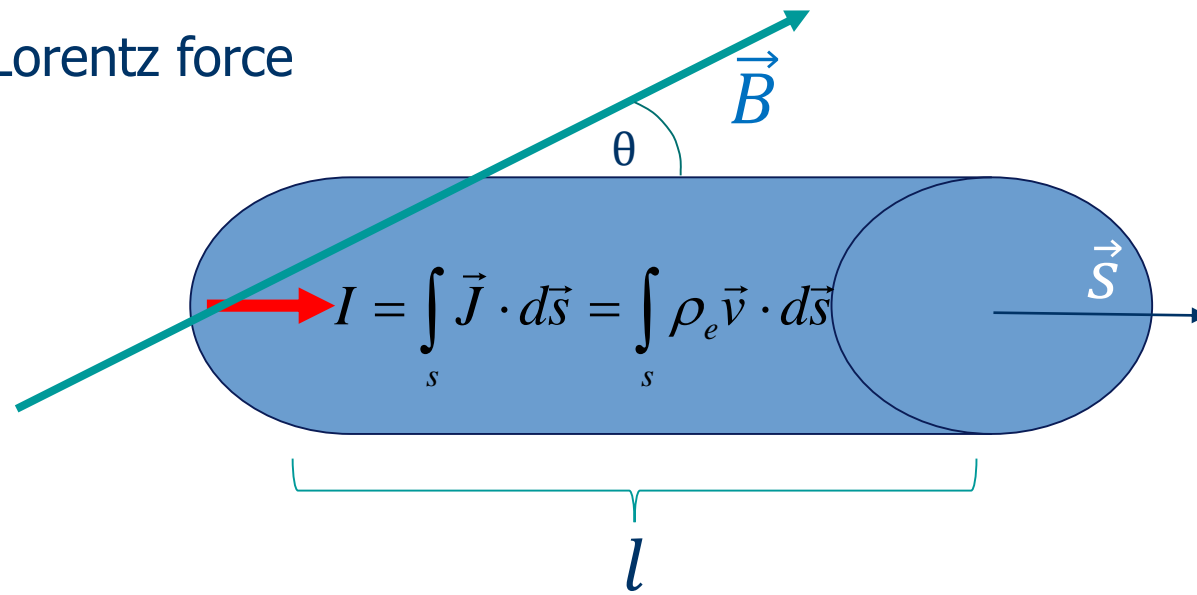
$$\oint_L \vec{B} \cdot d\vec{l} = \mu_0 I_f$$
$$\iint_S (\nabla \times \vec{B}) \cdot d\vec{s} = \mu_0 \iint_S \vec{J} \cdot d\vec{s} \Rightarrow \nabla \times \vec{B} = \mu_0 \vec{J}$$

Current is the source of magnetic field and charge is the source of electric field



Hendrik Antoon Lorentz 1853–1928

Lorentz force



The magnetic force acting upon a current carrying wire is given by:

$$|\vec{F}_B| = |\vec{I} \times \vec{B}| = IlB \sin \theta = JS l B \sin \theta = \rho_e v V B \sin \theta = Q v B \sin \theta = |Q \vec{v} \times \vec{B}|$$

And in the presence of electric field the total force is:

$$\vec{F} = Q(\vec{E} + \vec{v} \times \vec{B})$$

Faraday law



Michael Faraday (1791 – 1867)

$$V_{emf} = -\frac{d\phi_B}{dt} = -\frac{d}{dt} \int_s \vec{B} \cdot d\vec{s}$$

Hence the circulation integral should include the contribution of the magnetic field as well:

Conservative field

$$V_{Total} = \oint_l \vec{E}_T \cdot d\vec{l} = 0 + V_{emf} = -\frac{d\phi_B}{dt} = -\frac{d}{dt} \int_s \vec{B} \cdot d\vec{s} \Rightarrow \underbrace{\oint_l \vec{E}_T \cdot d\vec{l}} = -\frac{d}{dt} \int_s \vec{B} \cdot d\vec{s}$$

Stokes theorem

$$= \iint_s (\nabla \times \vec{E}) \cdot d\vec{s} = -\frac{d}{dt} \iint_s \vec{B} \cdot d\vec{s} \Rightarrow \nabla \times \vec{E} = -\frac{d\vec{B}}{dt}$$

Maxwell Equations



James Clerk Maxwell (1831–1879)

$$I : \oint_s \vec{E} \cdot d\vec{s} = \frac{q_{f,n}}{\epsilon_0} = \phi_e \quad ; \quad \text{div} \vec{E} = \frac{\rho_v}{\epsilon_0} \quad ; \quad \left[\oint_s \vec{D} \cdot d\vec{s} = q_{f,n} ; \text{div} \vec{D} = \rho_v \right] \quad ; \quad \vec{E} = -\text{grad} V ; \nabla^2 V = -\frac{\rho_v}{\epsilon_0}$$

$$II : \oint_s \vec{B} \cdot d\vec{s} = 0 \quad ; \quad \text{div} \vec{B} = 0 \quad ; \quad \left[\oint_s \vec{H} \cdot d\vec{s} = 0 ; \text{div} \vec{H} = 0 \right] \quad ; \quad \vec{B} = \text{curl} \vec{A} ; \nabla^2 \vec{A} = -\mu_0 \vec{J}$$

$$II : \oint_l \vec{E} \cdot d\vec{l} = -\frac{d\phi_B}{dt} = -\frac{d}{dt} \int_s \vec{B} \cdot d\vec{s} \quad ; \quad \nabla \times \vec{E} = -\frac{d\vec{B}}{dt} = -\frac{d(\nabla \times \vec{A})}{dt} = -\nabla \times \dot{\vec{A}} \Rightarrow \vec{E}_{\text{General}} = -\text{grad} V - \dot{\vec{A}}$$

$$IV : \oint_l \vec{B} \cdot d\vec{l} = \mu_0 I + ? \quad ; \quad \nabla \times \vec{B} = \mu_0 \vec{J} + ? \rightarrow$$

$$\rightarrow [\vec{H} \equiv \frac{\vec{B}}{\mu_0} \Rightarrow \oint_l \vec{H} \cdot d\vec{l} = I + ? \quad ; \quad \nabla \times \vec{H} = \vec{J} + ?]$$

Relating to the differential aspect of equation VI

$$\text{curl} \vec{H} = \vec{J} \rightarrow \underbrace{\text{div curl} \vec{H}}_{=0} = \underbrace{\text{div} \vec{J}}_{\neq 0}$$

Following the continuity law

$$\text{div} \vec{J} + \dot{\rho}_V = 0 \rightarrow \frac{\rho_V}{\epsilon_0} = \text{div} \vec{E} \rightarrow \dot{\rho}_V = \epsilon_0 \text{div} \dot{\vec{E}} \Rightarrow$$

$$\text{div}(\vec{J} + \epsilon_0 \dot{\vec{E}}) = 0 \quad \Rightarrow \quad \text{div}(\vec{J} + \dot{\vec{D}}) = \text{div}(\vec{J}_C + \vec{J}_D) = 0$$

$$I_D = \frac{d}{dt} \int_s \vec{D} \cdot d\vec{s} = \epsilon_0 \frac{d}{dt} \int_s \vec{E} \cdot d\vec{s} = \epsilon_0 \frac{d\phi_e}{dt}$$

$$\begin{aligned} IV : \oint_l \vec{B} \cdot d\vec{l} &= \mu_0(I_C + I_D) = \mu_0(I_C + \epsilon_0 \frac{d\phi_e}{dt}) = \mu_0(I_C + \epsilon_0 \frac{d}{dt} \int_s \vec{E} \cdot d\vec{s}) \Rightarrow \\ \Rightarrow \nabla \times \vec{B} &= \mu_0(\vec{J}_C + \vec{J}_D) = \mu_0(\vec{J}_C + \epsilon_0 \dot{\vec{E}}) = \mu_0(\vec{J}_C + \dot{\vec{D}}) \end{aligned}$$

The magnetic field intensity-H

$$\oint_L \vec{B} \cdot d\vec{l} = \mu_0 I_f \Rightarrow \oint_L \frac{\vec{B}}{\mu_0} \cdot d\vec{l} = I_f ; \quad \boxed{\frac{\vec{B}}{\mu_0} \equiv \vec{H} \Rightarrow \vec{B} = \mu_0 \vec{H}}$$

$$\Rightarrow \boxed{\oint_L \vec{H} \cdot d\vec{l} = I_f} \Rightarrow \{H\} = \frac{A}{m}$$

$$\boxed{\oint_L \vec{B} \cdot d\vec{l} = \mu_0 I_f} \rightarrow \text{Via Stokes Theorem: } \oint_L \vec{B} \cdot d\vec{l} = \iint_{S(L)} \nabla \times \vec{B} \cdot d\vec{s} = \iint_{S(L)} \mu_0 \vec{J}_f \cdot d\vec{s} \Rightarrow$$

$$\Rightarrow \boxed{\nabla \times \vec{B} = \mu_0 \vec{J}_f}$$

$$\boxed{\nabla \times \vec{H} = \vec{J}_f}$$

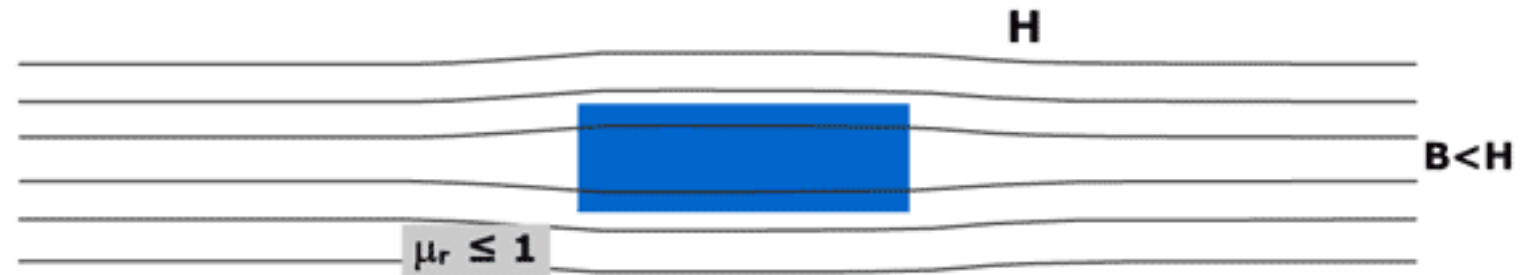
B : The magnetic flux density

H : The magnetic field intensity

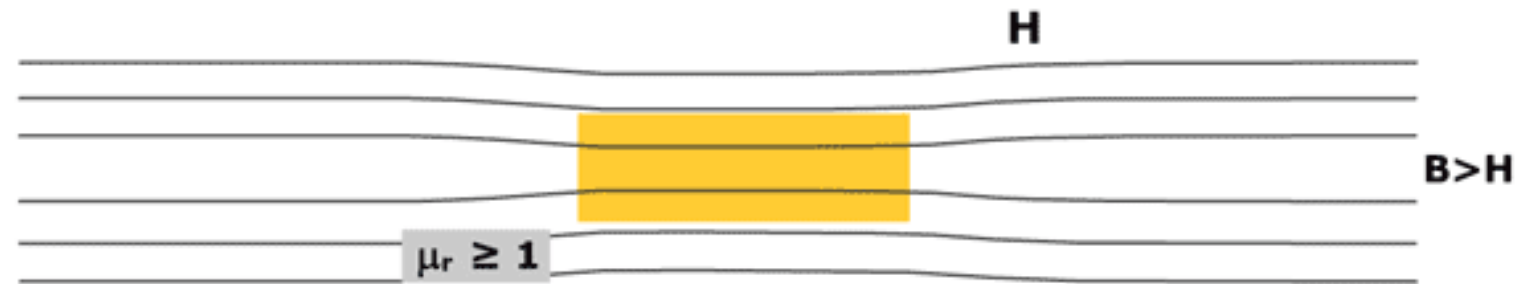
Types of magnetism:

Diamagnetism	Paramagnetism	Ferromagnetism	
No	Yes but weak	Yes and constant	Magnetic field
Electrons spin orbital	Electrons spin	Magnetizing regions	Source of magnetism
Opposite	Identical	Varies, Hysteresis	Direction of B in respect to H
Copper, Lead, Diamond, Mercury, Silicon	Aluminum, Calcium, Magnesium, Tungsten	Iron, Nickel, Cobalt	Typical materials
$\approx -10^{-5}$	$\approx 10^{-5}$	$ \chi_m \gg 1$	Typical χ_m
≈ 1	≈ 1	$ \mu_r \gg 1$	Typical μ_r

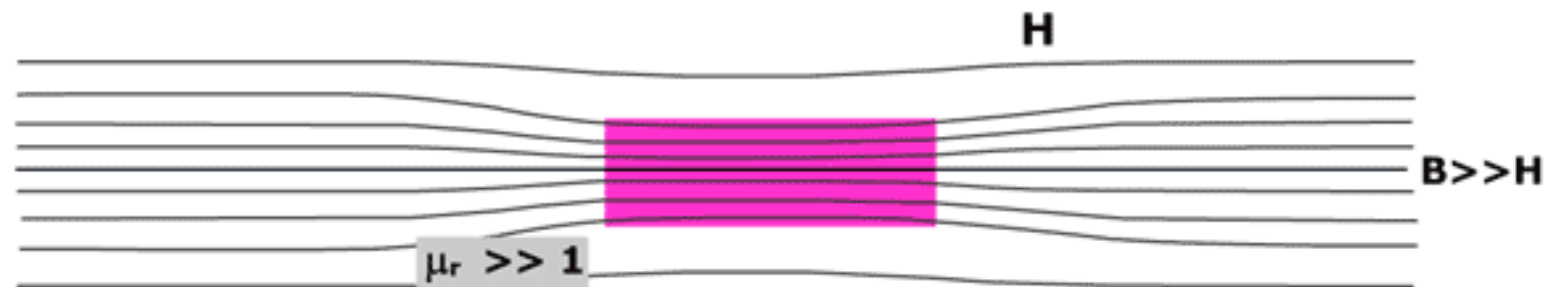
Diamagnetism

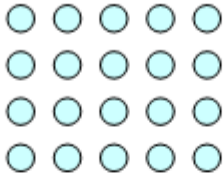
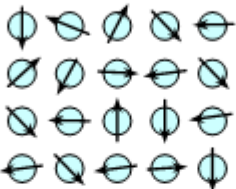
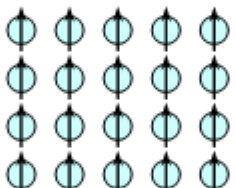
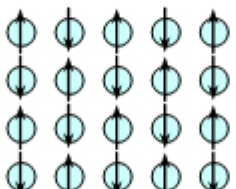
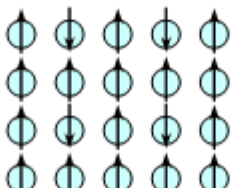


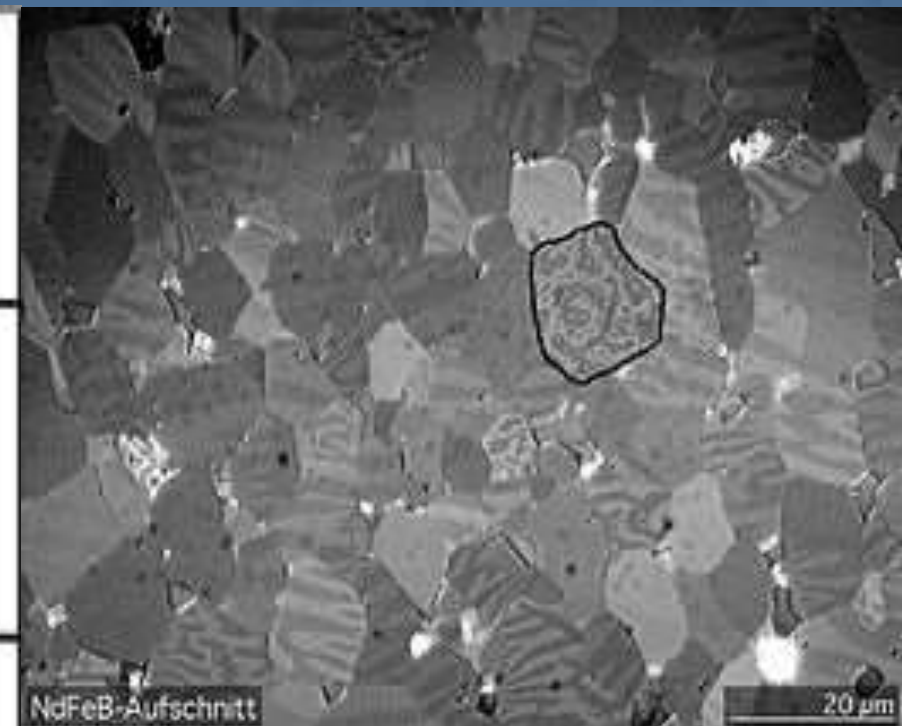
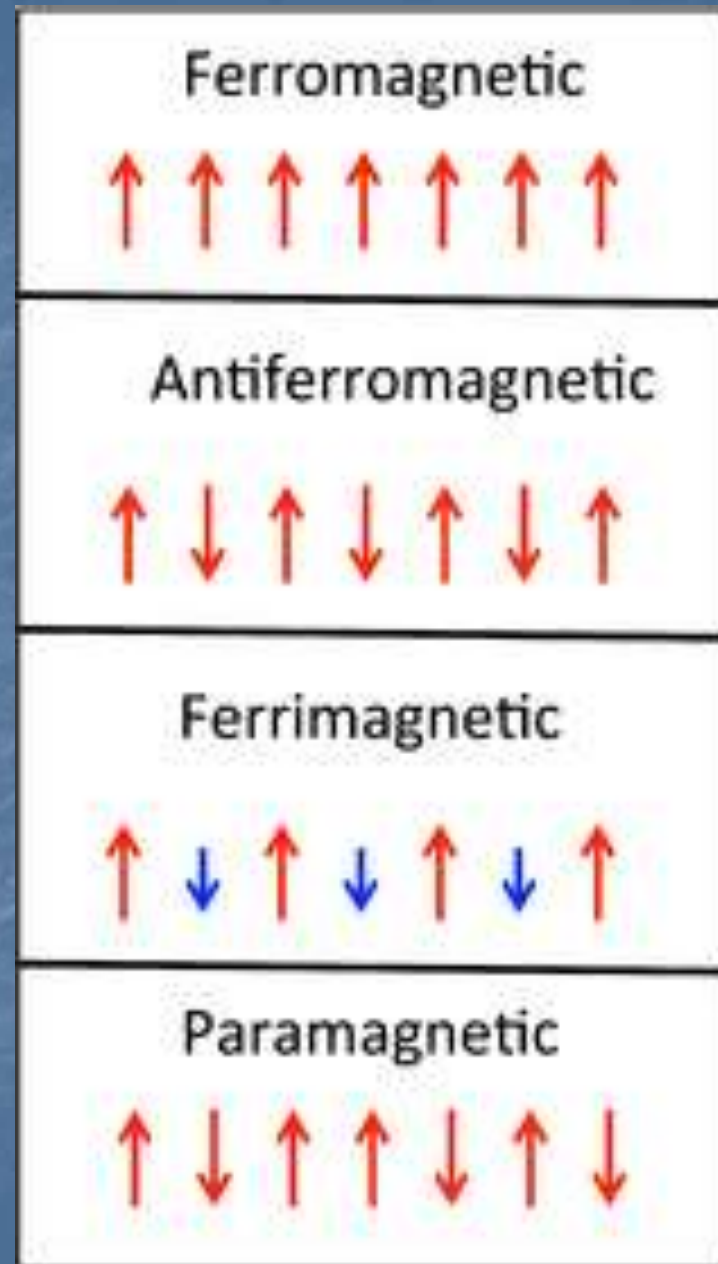
Paramagnetism



Ferromagnetism

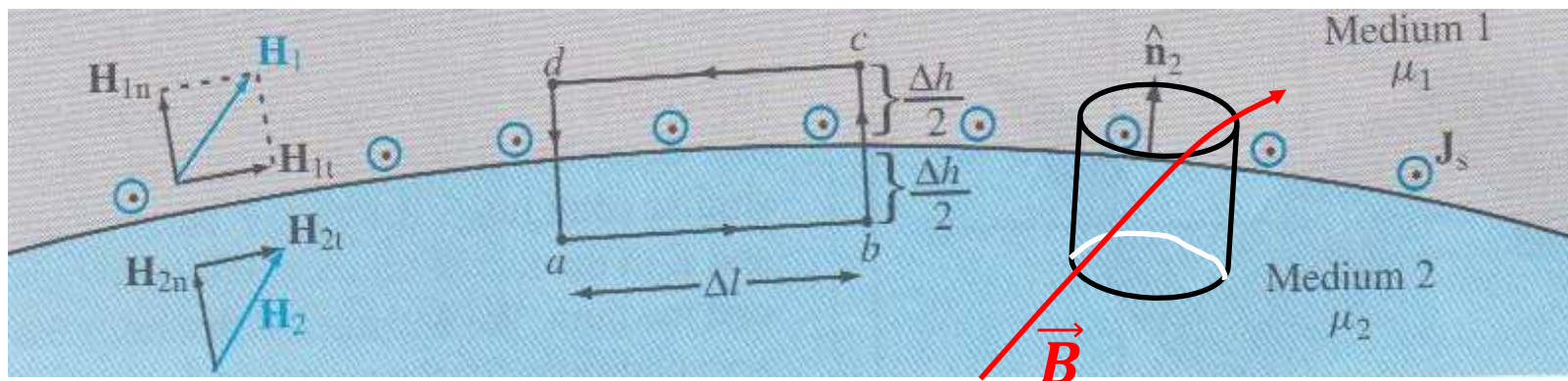


Type	Atomic / Magnetic Behaviour	
Dia-magnetism		Atoms have no magnetic moment
Para-magnetism		Atoms have randomly oriented magnetic moments
Ferro-magnetism		Atoms have parallel aligned magnetic moments
Antiferro-magnetism		Atoms have anti-parallel aligned magnetic moments
Ferri-magnetism		Atoms have mixed parallel and anti-parallel aligned magnetic moments



Within each grain are a series of lighter and darker stripes (imaged by using the optical [Kerr effect](#)) that are ferromagnetic domains with opposite orientations. Averaged over the whole sample, these domains have random orientation so the net magnetization is zero. [Moving domain](#) exist as well.

Boundary conditions for B and H



$$0 = \oint_S \vec{B} \cdot d\vec{s} \Rightarrow B_{N1} = B_{N2} \Rightarrow \mu_1 H_{N1} = \mu_2 H_{N2}$$

$$\oint_L \vec{H} \cdot d\vec{l} = I_f \Rightarrow (H_{T1} - H_{T2})L = I_f \Rightarrow (H_{T1} - H_{T2}) = \frac{I_f}{L} = i_f \Rightarrow \Delta H_T = i_f \Rightarrow$$

$$\Rightarrow \frac{B_{T1}}{\mu_1} - \frac{B_{T2}}{\mu_2} = i_f$$

For $i_f = 0$: $\Delta H_T = 0$ and $\frac{B_{T1}}{\mu_1} = \frac{B_{T2}}{\mu_2}$